

SELECTED FORMULAE

Coordinate transformations

$$\begin{aligned} \text{Cylindrical} \mapsto \text{rectangular} & \quad (x, y, z) = (r \cos \phi, r \sin \phi, z) \\ \text{Spherical} \mapsto \text{rectangular} & \quad (x, y, z) = (r \sin \theta \cos \phi, r \sin \theta \sin \phi, r \cos \theta) \end{aligned}$$

Differential operators

$$\begin{aligned} \text{Gradient} & \quad \nabla f = \frac{\partial f}{\partial x} \hat{\mathbf{x}} + \frac{\partial f}{\partial y} \hat{\mathbf{y}} + \frac{\partial f}{\partial z} \hat{\mathbf{z}} \\ \text{Divergence} & \quad \nabla \cdot \mathbf{F} = \frac{\partial F_x}{\partial x} + \frac{\partial F_y}{\partial y} + \frac{\partial F_z}{\partial z} \quad (\text{Rectangular}) \\ & \quad \nabla \cdot \mathbf{F} = \frac{1}{r} \frac{\partial}{\partial r} (r F_r) + \frac{1}{r} \frac{\partial F_\phi}{\partial \phi} + \frac{\partial F_z}{\partial z} \quad (\text{Cylindrical}) \\ \text{Curl} & \quad \nabla \times \mathbf{F} = \begin{vmatrix} \hat{\mathbf{x}} & \hat{\mathbf{y}} & \hat{\mathbf{z}} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ F_x & F_y & F_z \end{vmatrix} \end{aligned}$$

Line integral

$$\text{Path } \ell \quad \int_{\ell} \mathbf{F} \cdot d\mathbf{l} = \int_0^1 \mathbf{F}(\mathbf{l}(s)) \cdot \frac{d\mathbf{l}(s)}{ds} ds$$

Surface integrals

$$\begin{aligned} \text{Rectangular surface (} z = 0 \text{ plane)} & \quad \iint_A \mathbf{F} \cdot d\mathbf{A} = \int_a^b \int_c^d \mathbf{F}(x, y, 0) \cdot (dx dy \hat{\mathbf{z}}) \\ \text{Circular surface (} z = 0 \text{ plane)} & \quad \iint_A \mathbf{F} \cdot d\mathbf{A} = \int_0^R \int_0^{2\pi} \mathbf{F}(r, \phi, 0) \cdot (r d\phi dr \hat{\mathbf{z}}) \\ \text{Cylindrical surface (no end faces)} & \quad \iint_A \mathbf{F} \cdot d\mathbf{A} = \int_0^L \int_0^{2\pi} \mathbf{F}(R, \phi, z) \cdot (R d\phi dz \hat{\mathbf{r}}) \\ \text{Spherical surface} & \quad \iint_A \mathbf{F} \cdot d\mathbf{A} = \int_0^{2\pi} \int_0^\pi \mathbf{F}(R, \phi, \theta) \cdot (R^2 \sin \theta d\theta d\phi \hat{\mathbf{r}}) \end{aligned}$$

Volume integrals

$$\begin{aligned} \text{Cube} & \quad \iiint_v \rho dv = \int_0^a \int_0^a \int_0^a \rho(x, y, z) dx dy dz \\ \text{Cylinder} & \quad \iiint_v \rho dv = \int_0^L \int_0^R \int_0^{2\pi} \rho(r, \phi, z) r d\phi dr dz \\ \text{Sphere} & \quad \iiint_v \rho dv = \int_0^R \int_0^{2\pi} \int_0^\pi \rho(r, \phi, \theta) r^2 \sin \theta d\theta d\phi dr \end{aligned}$$

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Flux densities, fields and forces

Electric flux density and field	$\mathbf{D} = \epsilon \mathbf{E}$
Magnetic flux density and field	$\mathbf{B} = \mu \mathbf{H}$
Electric force and field	$\mathbf{F} = q \mathbf{E}$
Magnetic force and flux density	$\mathbf{F} = q \mathbf{v} \times \mathbf{B}$

Maxwell's Equations

Gauss (Electric)	$\nabla \cdot \mathbf{D} = \rho$	$\iint_A \mathbf{D} \cdot d\mathbf{A} = \iiint_{\text{vol}(A)} \rho dv$
Gauss (Magnetic)	$\nabla \cdot \mathbf{B} = 0$	$\iint_A \mathbf{B} \cdot d\mathbf{A} = 0$
Faraday	$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$	$\oint_{\ell(A)} \mathbf{E} \cdot d\mathbf{l} = -\frac{\partial}{\partial t} \iint_A \mathbf{B} \cdot d\mathbf{A}$
Ampere	$\nabla \times \mathbf{H} = \mathbf{J} + \frac{\partial \mathbf{D}}{\partial t}$	$\oint_{\ell(A)} \mathbf{H} \cdot d\mathbf{l} = \iint_A \mathbf{J} \cdot d\mathbf{A} + \frac{\partial}{\partial t} \iint_A \mathbf{D} \cdot d\mathbf{A}$

Conservation Laws

Energy	$\mathbf{E} = -\nabla V$	$V = -\int_{\ell} \mathbf{E} \cdot d\mathbf{l}$
Charge	$\nabla \cdot \mathbf{J} = -\frac{\partial \rho}{\partial t}$	$\iint_{\text{area}(v)} \mathbf{J} \cdot d\mathbf{A} = -\frac{\partial}{\partial t} \iiint_v \rho dv$

Conductors

Current	$i = \iint_A \mathbf{J} \cdot d\mathbf{A}$	
Voltage	$v = -\int_{\ell} \mathbf{E} \cdot d\mathbf{l}$	
Ohm's Law	$\mathbf{J} = \sigma \mathbf{E}$	$v = Ri$
Skin depth	$\delta = \sqrt{\frac{2}{\omega \mu \sigma}}$	
Permittivity (free space)	$\epsilon_o = 8.85 \times 10^{-12} \text{ F m}^{-1}$	
Permeability (free-space)	$\mu_o = 4\pi \times 10^{-7} \text{ H m}^{-1}$	
Conductivity (copper)	$\sigma = 5.8 \times 10^7 \text{ } (\Omega \text{ m})^{-1}$	
Permittivity (copper)	$\epsilon = \epsilon_o = 8.85 \times 10^{-12} \text{ F m}^{-1}$	
Permeability (copper)	$\mu = \mu_o = 4\pi \times 10^{-7} \text{ H m}^{-1}$	
Charge redistribution time	$T_{\text{CR}} = \frac{\epsilon}{\sigma}$	

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Conductors (continued)

Resistance

$$R = -\frac{\int_{\ell} \mathbf{E} \cdot d\mathbf{l}}{\iint_A \sigma \mathbf{E} \cdot d\mathbf{A}}$$

$$R = \int_0^L \frac{dx}{\sigma(x) A(x)}$$

$$R = \frac{L}{\sigma A}$$

Capacitance (linear)

$$C = -\frac{\epsilon \iiint_v \nabla \cdot \mathbf{E} dv}{\int_{\ell} \mathbf{E} \cdot d\mathbf{l}}$$

$$C = \frac{\epsilon A}{d}$$

Divergence in cylindrical coordinates

$$\nabla \cdot \mathbf{F} = \frac{1}{r} \frac{\partial}{\partial r} (r F_r) + \frac{1}{r} \frac{\partial F_{\phi}}{\partial \phi} + \frac{\partial F_z}{\partial z}$$

A possibly useful formula

$$\frac{1}{b^2 - z^2} = \frac{1}{2b} \left[\frac{1}{b - z} + \frac{1}{b + z} \right]$$

Conversion between Cartesian, cylindrical, and spherical coordinates

		From		
		Cartesian	Cylindrical	Spherical
To	Cartesian	N/A	$x = \rho \cos \varphi$ $y = \rho \sin \varphi$ $z = z$	$x = r \sin \theta \cos \varphi$ $y = r \sin \theta \sin \varphi$ $z = r \cos \theta$
	Cylindrical	$\rho = \sqrt{x^2 + y^2}$ $\varphi = \arctan\left(\frac{y}{x}\right)$ $z = z$	N/A	$\rho = r \sin \theta$ $\varphi = \varphi$ $z = r \cos \theta$
	Spherical	$r = \sqrt{x^2 + y^2 + z^2}$ $\theta = \arccos\left(\frac{z}{r}\right)$ $\varphi = \arctan\left(\frac{y}{x}\right)$	$r = \sqrt{\rho^2 + z^2}$ $\theta = \arctan\left(\frac{\rho}{z}\right)$ $\varphi = \varphi$	N/A

Conversion between unit vectors in Cartesian, cylindrical, and spherical coordinate systems in terms of destination coordinates

	Cartesian	Cylindrical	Spherical
Cartesian	N/A	$\hat{x} = \cos \varphi \hat{\rho} - \sin \varphi \hat{\varphi}$ $\hat{y} = \sin \varphi \hat{\rho} + \cos \varphi \hat{\varphi}$ $\hat{z} = \hat{z}$	$\hat{x} = \sin \theta \cos \varphi \hat{r} + \cos \theta \cos \varphi \hat{\theta} - \sin \varphi \hat{\varphi}$ $\hat{y} = \sin \theta \sin \varphi \hat{r} + \cos \theta \sin \varphi \hat{\theta} + \cos \varphi \hat{\varphi}$ $\hat{z} = \cos \theta \hat{r} - \sin \theta \hat{\theta}$
Cylindrical	$\hat{\rho} = \frac{x\hat{x} + y\hat{y}}{\sqrt{x^2 + y^2}}$ $\hat{\varphi} = \frac{-y\hat{x} + x\hat{y}}{\sqrt{x^2 + y^2}}$ $\hat{z} = \hat{z}$	N/A	$\hat{\rho} = \sin \theta \hat{r} + \cos \theta \hat{\theta}$ $\hat{\varphi} = \hat{\varphi}$ $\hat{z} = \cos \theta \hat{r} - \sin \theta \hat{\theta}$
Spherical	$\hat{r} = \frac{x\hat{x} + y\hat{y} + z\hat{z}}{\sqrt{x^2 + y^2 + z^2}}$ $\hat{\theta} = \frac{(x\hat{x} + y\hat{y})z - (x^2 + y^2)\hat{z}}{\sqrt{x^2 + y^2 + z^2}\sqrt{x^2 + y^2}}$ $\hat{\varphi} = \frac{-y\hat{x} + x\hat{y}}{\sqrt{x^2 + y^2}}$	$\hat{r} = \frac{\rho\hat{\rho} + z\hat{z}}{\sqrt{\rho^2 + z^2}}$ $\hat{\theta} = \frac{z\hat{\rho} - \rho\hat{z}}{\sqrt{\rho^2 + z^2}}$ $\hat{\varphi} = \hat{\varphi}$	N/A

Conversion between unit vectors in Cartesian, cylindrical, and spherical coordinate systems in terms of source coordinates

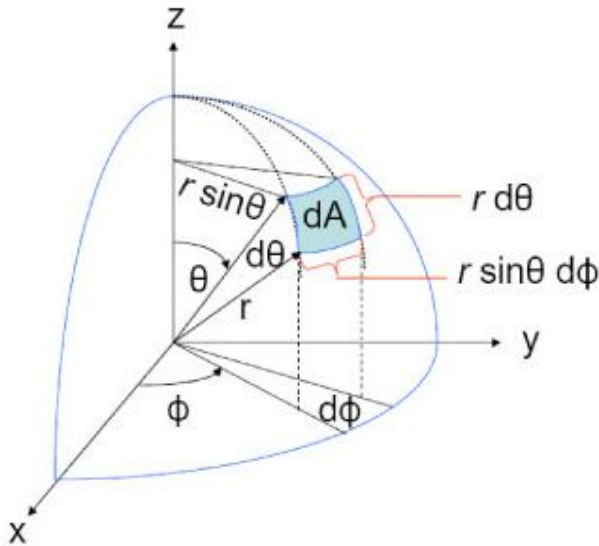
	Cartesian	Cylindrical	Spherical
Cartesian	N/A	$\hat{x} = \frac{x\hat{\rho} - y\hat{\varphi}}{\sqrt{x^2 + y^2}}$ $\hat{y} = \frac{y\hat{\rho} + x\hat{\varphi}}{\sqrt{x^2 + y^2}}$ $\hat{z} = \hat{z}$	$\hat{x} = \frac{x(\sqrt{x^2 + y^2}\hat{r} + z\hat{\theta}) - y\sqrt{x^2 + y^2 + z^2}\hat{\varphi}}{\sqrt{x^2 + y^2}\sqrt{x^2 + y^2 + z^2}}$ $\hat{y} = \frac{y(\sqrt{x^2 + y^2}\hat{r} + z\hat{\theta}) + x\sqrt{x^2 + y^2 + z^2}\hat{\varphi}}{\sqrt{x^2 + y^2}\sqrt{x^2 + y^2 + z^2}}$ $\hat{z} = \frac{z\hat{r} - \sqrt{x^2 + y^2}\hat{\theta}}{\sqrt{x^2 + y^2 + z^2}}$
Cylindrical	$\hat{\rho} = \cos \varphi \hat{x} + \sin \varphi \hat{y}$ $\hat{\varphi} = -\sin \varphi \hat{x} + \cos \varphi \hat{y}$ $\hat{z} = \hat{z}$	N/A	$\hat{\rho} = \frac{\rho\hat{r} + z\hat{\theta}}{\sqrt{\rho^2 + z^2}}$ $\hat{\varphi} = \hat{\varphi}$ $\hat{z} = \frac{z\hat{r} - \rho\hat{\theta}}{\sqrt{\rho^2 + z^2}}$
Spherical	$\hat{r} = \sin \theta (\cos \varphi \hat{x} + \sin \varphi \hat{y}) + \cos \theta \hat{z}$ $\hat{\theta} = \cos \theta (\cos \varphi \hat{x} + \sin \varphi \hat{y}) - \sin \theta \hat{z}$ $\hat{\varphi} = -\sin \varphi \hat{x} + \cos \varphi \hat{y}$	$\hat{r} = \sin \theta \hat{\rho} + \cos \theta \hat{z}$ $\hat{\theta} = \cos \theta \hat{\rho} - \sin \theta \hat{z}$ $\hat{\varphi} = \hat{\varphi}$	N/A

Table with the del operator in cartesian, cylindrical and spherical coordinates

Operation	Cartesian coordinates (x, y, z)	Cylindrical coordinates (ρ, φ, z)	Spherical coordinates (r, θ, φ), where θ is the polar angle and φ is azimuthal
A vector field A	$A_x \hat{x} + A_y \hat{y} + A_z \hat{z}$	$A_\rho \hat{\rho} + A_\varphi \hat{\varphi} + A_z \hat{z}$	$A_r \hat{r} + A_\theta \hat{\theta} + A_\varphi \hat{\varphi}$
Gradient ∇f	$\frac{\partial f}{\partial x} \hat{x} + \frac{\partial f}{\partial y} \hat{y} + \frac{\partial f}{\partial z} \hat{z}$	$\frac{\partial f}{\partial \rho} \hat{\rho} + \frac{1}{\rho} \frac{\partial f}{\partial \varphi} \hat{\varphi} + \frac{\partial f}{\partial z} \hat{z}$	$\frac{\partial f}{\partial r} \hat{r} + \frac{1}{r} \frac{\partial f}{\partial \theta} \hat{\theta} + \frac{1}{r \sin \theta} \frac{\partial f}{\partial \varphi} \hat{\varphi}$
Divergence $\nabla \cdot \mathbf{A}$	$\frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z}$	$\frac{1}{\rho} \frac{\partial(\rho A_\rho)}{\partial \rho} + \frac{1}{\rho} \frac{\partial A_\varphi}{\partial \varphi} + \frac{\partial A_z}{\partial z}$	$\frac{1}{r^2} \frac{\partial(r^2 A_r)}{\partial r} + \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (A_\theta \sin \theta) + \frac{1}{r \sin \theta} \frac{\partial A_\varphi}{\partial \varphi}$
Curl $\nabla \times \mathbf{A}$	$\left(\frac{\partial A_z}{\partial y} - \frac{\partial A_y}{\partial z} \right) \hat{x}$ $+ \left(\frac{\partial A_x}{\partial z} - \frac{\partial A_z}{\partial x} \right) \hat{y}$ $+ \left(\frac{\partial A_y}{\partial x} - \frac{\partial A_x}{\partial y} \right) \hat{z}$	$\left(\frac{1}{\rho} \frac{\partial A_z}{\partial \varphi} - \frac{\partial A_\varphi}{\partial z} \right) \hat{\rho}$ $+ \left(\frac{\partial A_\rho}{\partial z} - \frac{\partial A_z}{\partial \rho} \right) \hat{\varphi}$ $+ \frac{1}{\rho} \left(\frac{\partial(\rho A_\varphi)}{\partial \rho} - \frac{\partial A_\rho}{\partial \varphi} \right) \hat{z}$	$\frac{1}{r \sin \theta} \left(\frac{\partial}{\partial \theta} (A_\varphi \sin \theta) - \frac{\partial A_\theta}{\partial \varphi} \right) \hat{r}$ $+ \frac{1}{r} \left(\frac{1}{\sin \theta} \frac{\partial A_r}{\partial \varphi} - \frac{\partial}{\partial r} (r A_\varphi) \right) \hat{\theta}$ $+ \frac{1}{r} \left(\frac{\partial}{\partial r} (r A_\theta) - \frac{\partial A_r}{\partial \theta} \right) \hat{\varphi}$
Laplace operator $\nabla^2 f \equiv \Delta f$	$\frac{\partial^2 f}{\partial x^2} + \frac{\partial^2 f}{\partial y^2} + \frac{\partial^2 f}{\partial z^2}$	$\frac{1}{\rho} \frac{\partial}{\partial \rho} \left(\rho \frac{\partial f}{\partial \rho} \right) + \frac{1}{\rho^2} \frac{\partial^2 f}{\partial \varphi^2} + \frac{\partial^2 f}{\partial z^2}$	$\frac{1}{r^2} \frac{\partial}{\partial r} \left(r^2 \frac{\partial f}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left(\sin \theta \frac{\partial f}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 f}{\partial \varphi^2}$
Vector Laplacian $\nabla^2 \mathbf{A} \equiv \Delta \mathbf{A}$	$\nabla^2 A_x \hat{x} + \nabla^2 A_y \hat{y} + \nabla^2 A_z \hat{z}$	— View by clicking [show] — [show]	— View by clicking [show] — [show]
Material derivative ^[1] $(\mathbf{A} \cdot \nabla) \mathbf{B}$	$\mathbf{A} \cdot \nabla B_x \hat{x} + \mathbf{A} \cdot \nabla B_y \hat{y} + \mathbf{A} \cdot \nabla B_z \hat{z}$	— View by clicking [show] — [show]	— View by clicking [show] — [show]
Tensor divergence $\nabla \cdot \mathbf{T}$	— View by clicking [show] — [show]	— View by clicking [show] — [show]	— View by clicking [show] — [show]
Differential displacement $d\mathbf{l}$	$dx \hat{x} + dy \hat{y} + dz \hat{z}$	$d\rho \hat{\rho} + \rho d\varphi \hat{\varphi} + dz \hat{z}$	$dr \hat{r} + r d\theta \hat{\theta} + r \sin \theta d\varphi \hat{\varphi}$
Differential normal area $d\mathbf{S}$	$dy dz \hat{x}$ $+ dx dz \hat{y}$ $+ dx dy \hat{z}$	$\rho d\varphi dz \hat{\rho}$ $+ d\rho dz \hat{\varphi}$ $+ \rho d\rho d\varphi \hat{z}$	$r^2 \sin \theta d\theta d\varphi \hat{r}$ $+ r \sin \theta dr d\varphi \hat{\theta}$ $+ r dr d\theta \hat{\varphi}$
Differential volume dV	$dx dy dz$	$\rho d\rho d\varphi dz$	$r^2 \sin \theta dr d\theta d\varphi$

Non-trivial calculation rules [\[edit\]](#)

1. $\text{div grad } f \equiv \nabla \cdot \nabla f \equiv \nabla^2 f$
2. $\text{curl grad } f \equiv \nabla \times \nabla f = \mathbf{0}$
3. $\text{div curl } \mathbf{A} \equiv \nabla \cdot (\nabla \times \mathbf{A}) = 0$
4. $\text{curl curl } \mathbf{A} \equiv \nabla \times (\nabla \times \mathbf{A}) = \nabla(\nabla \cdot \mathbf{A}) - \nabla^2 \mathbf{A}$ (Lagrange's fomula for del)
5. $\nabla^2(fg) = f\nabla^2g + 2\nabla f \cdot \nabla g + g\nabla^2f$



$$dA = \sqrt{r^4 \sin^2(\theta)} d\theta d\varphi = r^2 \sin(\theta) d\theta d\varphi$$

Formulas and Identities

Tangent and Cotangent Identities

$$\tan \theta = \frac{\sin \theta}{\cos \theta} \qquad \cot \theta = \frac{\cos \theta}{\sin \theta}$$

Reciprocal Identities

$$\csc \theta = \frac{1}{\sin \theta} \qquad \sin \theta = \frac{1}{\csc \theta}$$

$$\sec \theta = \frac{1}{\cos \theta} \qquad \cos \theta = \frac{1}{\sec \theta}$$

$$\cot \theta = \frac{1}{\tan \theta} \qquad \tan \theta = \frac{1}{\cot \theta}$$

Pythagorean Identities

$$\sin^2 \theta + \cos^2 \theta = 1$$

$$\tan^2 \theta + 1 = \sec^2 \theta$$

$$1 + \cot^2 \theta = \csc^2 \theta$$

Even/Odd Formulas

$$\sin(-\theta) = -\sin \theta \qquad \csc(-\theta) = -\csc \theta$$

$$\cos(-\theta) = \cos \theta \qquad \sec(-\theta) = \sec \theta$$

$$\tan(-\theta) = -\tan \theta \qquad \cot(-\theta) = -\cot \theta$$

Periodic Formulas

If n is an integer.

$$\sin(\theta + 2\pi n) = \sin \theta \qquad \csc(\theta + 2\pi n) = \csc \theta$$

$$\cos(\theta + 2\pi n) = \cos \theta \qquad \sec(\theta + 2\pi n) = \sec \theta$$

$$\tan(\theta + \pi n) = \tan \theta \qquad \cot(\theta + \pi n) = \cot \theta$$

Double Angle Formulas

$$\sin(2\theta) = 2 \sin \theta \cos \theta$$

$$\cos(2\theta) = \cos^2 \theta - \sin^2 \theta$$

$$= 2 \cos^2 \theta - 1$$

$$= 1 - 2 \sin^2 \theta$$

$$\tan(2\theta) = \frac{2 \tan \theta}{1 - \tan^2 \theta}$$

Degrees to Radians Formulas

If x is an angle in degrees and t is an angle in radians then

$$\frac{\pi}{180} = \frac{t}{x} \quad \Rightarrow \quad t = \frac{\pi x}{180} \quad \text{and} \quad x = \frac{180t}{\pi}$$

Half Angle Formulas (alternate form)

$$\sin \frac{\theta}{2} = \pm \sqrt{\frac{1 - \cos \theta}{2}} \qquad \sin^2 \theta = \frac{1}{2}(1 - \cos(2\theta))$$

$$\cos \frac{\theta}{2} = \pm \sqrt{\frac{1 + \cos \theta}{2}} \qquad \cos^2 \theta = \frac{1}{2}(1 + \cos(2\theta))$$

$$\tan \frac{\theta}{2} = \pm \sqrt{\frac{1 - \cos \theta}{1 + \cos \theta}} \qquad \tan^2 \theta = \frac{1 - \cos(2\theta)}{1 + \cos(2\theta)}$$

Sum and Difference Formulas

$$\sin(\alpha \pm \beta) = \sin \alpha \cos \beta \pm \cos \alpha \sin \beta$$

$$\cos(\alpha \pm \beta) = \cos \alpha \cos \beta \mp \sin \alpha \sin \beta$$

$$\tan(\alpha \pm \beta) = \frac{\tan \alpha \pm \tan \beta}{1 \mp \tan \alpha \tan \beta}$$

Product to Sum Formulas

$$\sin \alpha \sin \beta = \frac{1}{2}[\cos(\alpha - \beta) - \cos(\alpha + \beta)]$$

$$\cos \alpha \cos \beta = \frac{1}{2}[\cos(\alpha - \beta) + \cos(\alpha + \beta)]$$

$$\sin \alpha \cos \beta = \frac{1}{2}[\sin(\alpha + \beta) + \sin(\alpha - \beta)]$$

$$\cos \alpha \sin \beta = \frac{1}{2}[\sin(\alpha + \beta) - \sin(\alpha - \beta)]$$

Sum to Product Formulas

$$\sin \alpha + \sin \beta = 2 \sin\left(\frac{\alpha + \beta}{2}\right) \cos\left(\frac{\alpha - \beta}{2}\right)$$

$$\sin \alpha - \sin \beta = 2 \cos\left(\frac{\alpha + \beta}{2}\right) \sin\left(\frac{\alpha - \beta}{2}\right)$$

$$\cos \alpha + \cos \beta = 2 \cos\left(\frac{\alpha + \beta}{2}\right) \cos\left(\frac{\alpha - \beta}{2}\right)$$

$$\cos \alpha - \cos \beta = -2 \sin\left(\frac{\alpha + \beta}{2}\right) \sin\left(\frac{\alpha - \beta}{2}\right)$$

Cofunction Formulas

$$\sin\left(\frac{\pi}{2} - \theta\right) = \cos \theta \qquad \cos\left(\frac{\pi}{2} - \theta\right) = \sin \theta$$

$$\csc\left(\frac{\pi}{2} - \theta\right) = \sec \theta \qquad \sec\left(\frac{\pi}{2} - \theta\right) = \csc \theta$$

$$\tan\left(\frac{\pi}{2} - \theta\right) = \cot \theta \qquad \cot\left(\frac{\pi}{2} - \theta\right) = \tan \theta$$

Math Formulas: Hyperbolic functions

Definitions of hyperbolic functions

1. $\sinh x = \frac{e^x - e^{-x}}{2}$
2. $\cosh x = \frac{e^x + e^{-x}}{2}$
3. $\tanh x = \frac{e^x - e^{-x}}{e^x + e^{-x}} = \frac{\sinh x}{\cosh x}$
4. $\operatorname{csch} x = \frac{2}{e^x - e^{-x}} = \frac{1}{\sinh x}$
5. $\operatorname{sech} x = \frac{2}{e^x + e^{-x}} = \frac{1}{\cosh x}$
6. $\operatorname{coth} x = \frac{e^x + e^{-x}}{e^x - e^{-x}} = \frac{\cosh x}{\sinh x}$

Derivatives

7. $\frac{d}{dx} \sinh x = \cosh x$
8. $\frac{d}{dx} \cosh x = \sinh x$
9. $\frac{d}{dx} \tanh x = \operatorname{sech}^2 x$
10. $\frac{d}{dx} \operatorname{csch} x = -\operatorname{csch} x \cdot \operatorname{coth} x$
11. $\frac{d}{dx} \operatorname{sech} x = -\operatorname{sech} x \cdot \tanh x$
12. $\frac{d}{dx} \operatorname{coth} x = -\operatorname{csch}^2 x$

Hyperbolic identities

13. $\cosh^2 x - \sinh^2 x = 1$
14. $\tanh^2 x + \operatorname{sech}^2 x = 1$
15. $\operatorname{coth}^2 x - \operatorname{csch}^2 x = 1$
16. $\sinh(x \pm y) = \sinh x \cdot \cosh y \pm \cosh x \cdot \sinh y$
17. $\cosh(x \pm y) = \cosh x \cdot \cosh y \pm \sinh x \cdot \sinh y$
18. $\sinh(2 \cdot x) = 2 \cdot \sinh x \cdot \cosh x$
19. $\cosh(2 \cdot x) = \cosh^2 x + \sinh^2 x$

$$20. \quad \sinh^2 x = \frac{-1 + \cosh 2x}{2}$$

$$21. \quad \cosh^2 x = \frac{1 + \cosh 2x}{2}$$

Inverse Hyperbolic functions

$$22. \quad \sinh^{-1} x = \ln \left(x + \sqrt{x^2 + 1} \right), \quad x \in (-\infty, \infty)$$

$$23. \quad \cosh^{-1} x = \ln \left(x + \sqrt{x^2 - 1} \right), \quad x \in [1, \infty)$$

$$24. \quad \tanh^{-1} x = \frac{1}{2} \ln \left(\frac{1+x}{1-x} \right), \quad x \in (-1, 1)$$

$$25. \quad \coth^{-1} x = \frac{1}{2} \ln \left(\frac{x+1}{x-1} \right), \quad x \in (-\infty, -1) \cup (1, \infty)$$

$$26. \quad \operatorname{sech}^{-1} x = \ln \left(\frac{1 + \sqrt{1 - x^2}}{x} \right), \quad x \in (0, 1]$$

$$27. \quad \operatorname{csch}^{-1} x = \ln \left(\frac{1}{x} + \frac{\sqrt{1 - x^2}}{|x|} \right), \quad x \in (-\infty, 0) \cup (0, \infty)$$

Derivatives of Inverse Hyperbolic functions

$$28. \quad \frac{d}{dx} \sinh^{-1} x = \frac{1}{\sqrt{x^2 + 1}}$$

$$29. \quad \frac{d}{dx} \cosh^{-1} x = \frac{1}{\sqrt{x^2 - 1}}$$

$$30. \quad \frac{d}{dx} \tanh^{-1} x = \frac{1}{1 - x^2}$$

$$31. \quad \frac{d}{dx} \operatorname{csch}^{-1} x = -\frac{1}{|x|\sqrt{1 + x^2}}$$

$$32. \quad \frac{d}{dx} \operatorname{sech}^{-1} x = -\frac{1}{x\sqrt{1 - x^2}}$$

$$33. \quad \frac{d}{dx} \coth^{-1} x = \frac{1}{1 - x^2}$$

Formulas from Calculus

Derivatives

$$\begin{array}{llllll}
 \frac{d}{dx} [x^n] & = & nx^{n-1} & \frac{d}{dx} [e^x] & = & e^x & \frac{d}{dx} [\sin x] & = & \cos x \\
 \frac{d}{dx} [c] & = & 0 & \frac{d}{dx} [b^x] & = & b^x \ln b & \frac{d}{dx} [\cos x] & = & -\sin x \\
 \frac{d}{dx} [x] & = & 1 & \frac{d}{dx} [\ln x] & = & \frac{1}{x} & \frac{d}{dx} [\tan x] & = & \sec^2 x \\
 \frac{d}{dx} \left[\frac{1}{x} \right] & = & -\frac{1}{x^2} & \frac{d}{dx} [\log_b x] & = & \frac{1}{x \ln b} & \frac{d}{dx} [\sec x] & = & \tan x \sec x \\
 \frac{d}{dx} \left[\frac{1}{x^2} \right] & = & -\frac{2}{x^3} & \frac{d}{dx} [\sinh x] & = & \cosh x & \frac{d}{dx} [\arcsin x] & = & \frac{1}{\sqrt{1-x^2}} \\
 \frac{d}{dx} [\sqrt{x}] & = & \frac{1}{2\sqrt{x}} & \frac{d}{dx} [\cosh x] & = & \sinh x & \frac{d}{dx} [\arctan x] & = & \frac{1}{1+x^2} \\
 \frac{d}{dx} \left[\frac{1}{\sqrt{x}} \right] & = & -\frac{1}{2x\sqrt{x}} & \frac{d}{dx} [\tanh x] & = & \operatorname{sech}^2 x \\
 & & & \frac{d}{dx} [\operatorname{arcsinh} x] & = & \frac{1}{\sqrt{1+x^2}} \\
 & & & \frac{d}{dx} [\operatorname{arctanh} x] & = & \frac{1}{1-x^2}
 \end{array}$$

Product Rule: $\frac{d}{dx} [f(x)g(x)] = f'(x)g(x) + f(x)g'(x)$

Quotient Rule: $\frac{d}{dx} \left[\frac{f(x)}{g(x)} \right] = \frac{g(x)f'(x) - f(x)g'(x)}{g(x)^2}$

Chain Rule: $\frac{d}{dx} [f(g(x))] = f'(g(x))g'(x)$ or $\frac{dy}{dx} = \frac{dy}{du} \frac{du}{dx}$

Special Cases

$$\frac{d}{dx} [f(x)^n] = nf(x)^{n-1} f'(x)$$

$$\frac{d}{dx} \left[\frac{1}{g(x)} \right] = \frac{-g'(x)}{g(x)^2}$$

$$\frac{d}{dx} [\ln |f(x)|] = \frac{f'(x)}{f(x)}$$

$$\frac{d}{dx} [e^{f(x)}] = f'(x)e^{f(x)}$$

Integrals

$$\begin{array}{ll}
 \int x^n dx & = \frac{1}{n+1}x^{n+1} + C & \int e^x dx & = e^x + C \\
 \int \frac{1}{x} dx & = \ln|x| + C & \int b^x dx & = \frac{1}{\ln b}b^x \\
 \int c dx & = cx + C & \int \sinh x dx & = \cosh x + C \\
 \int x dx & = \frac{1}{2}x^2 + C & \int \cosh x dx & = \sinh x + C \\
 \int x^2 dx & = \frac{1}{3}x^3 + C & \int \sin x dx & = -\cos x + C \\
 \int \frac{1}{x^2} dx & = -\frac{1}{x} + C & \int \cos x dx & = \sin x + C \\
 \int \sqrt{x} dx & = \frac{2}{3}x\sqrt{x} + C & \int \tan x dx & = \ln|\sec x| + C \\
 \int \frac{1}{\sqrt{x}} dx & = 2\sqrt{x} + C & \int \sec x dx & = \ln|\tan x + \sec x| + C \\
 \int \frac{1}{1+x^2} dx & = \arctan x + C & \int \sin^2 x dx & = \frac{1}{2}(x - \sin x \cos x) + C \\
 \int \frac{1}{\sqrt{1-x^2}} dx & = \arcsin x + C & \int \cos^2 x dx & = \frac{1}{2}(x + \sin x \cos x) + C \\
 \int \ln x dx & = x \ln x - x + C & \int \tan^2 x dx & = \tan x - x + C \\
 \int x^n \ln x dx & = \frac{x^{n+1}}{n+1} \ln x - \frac{x^{n+1}}{(n+1)^2} + C & \int \sec^2 x dx & = \tan x + C
 \end{array}$$

Substitution $\int f(g(x))g'(x) dx = \int f(u) du = F(u) + C = F(g(x)) + C$

$$\int_a^b f(g(x))g'(x) dx = \int_{g(a)}^{g(b)} f(u) du$$

Special cases $\int \frac{f'(x)}{f(x)} dx = \ln|f(x)| + C$

$$\int e^{f(x)} f'(x) dx = e^{f(x)} + C$$

By parts $\int u dv = uv - \int v du$

$$\int_a^b u dv = uv]_a^b - \int_a^b v du$$

or $\int f(x)g'(x)dx = f(x)g(x) - \int g(x)f'(x) dx$

$$\int_a^b f(x)g'(x)dx = f(x)g(x)]_a^b - \int_a^b g(x)f'(x) dx$$

1 Resistors

\mathbf{J}		Current Density
\mathbf{v}		Velocity
\mathbf{E}		Electric Field
σ		Conductivity
ρ		Resistivity
μ		Permeability
q		Charge per carrier
n		Carrier density
m		Carrier mass
τ		Time between collisions
μ_q		Mobility (ease with which a charge carrier can drift)

Put above two together to get vector form of Ohm's Law

$$\mathbf{J} = \sigma \mathbf{E}$$

$$\sigma = \frac{q^2 \tau n}{m}$$

Skin Depth, the effective cross sectional area

$$\delta = \sqrt{\frac{2}{\omega \mu \sigma}}$$

Resistance of conductor skin

$$R(\delta) = \frac{l}{\sigma A(\delta)}$$

Equation for mobility

$$\mu_q = \frac{q\tau}{m}$$

Conductivity (dependent on charge, concentration and mobility)

$$\sigma = qn\mu_q$$

Essentially $R=V/I$

$$R = -\frac{\int_l \mathbf{E} \cdot d\mathbf{l}}{\iint_A \sigma \mathbf{E} \cdot d\mathbf{A}}$$

Volume integration over resistor

$$R = \int_0^L \frac{dx}{\sigma(x)A(x)}$$

Simplified version of volume integration for constant values

$$R = \frac{L}{\sigma A} = \frac{\rho L}{A}$$

Current density due to charge motion

$$\mathbf{J} = qn\mathbf{v}$$

Average charge velocity in resistor accounting for collisions

$$\mathbf{v} = \left(\frac{q\tau}{m}\right)\mathbf{E}$$

2 Conductors and Insulators

ϵ | Permittivity

σ | Conductivity

Low frequency behaviour is dominated by σ , high frequency behaviour determined by how long charge takes to redistribute in the material, which is a material dependent property

$$T_{CR} = \frac{\epsilon}{\sigma}$$

Charge redistribution bandwidth is the corner frequency between being a conductor and insulator

$$B_{CR} = \frac{1}{2\pi T_{CR}} = \frac{\sigma}{2\pi\epsilon}$$

$f \ll B_{CR}$ | Conductor

$f \gg B_{CR}$ | Insulator

Skin depth model used when it is significant, large conductivity means large current that generates charge separation which creates opposing electric field that attenuates the original field

$$\omega \ll \omega_{CR}$$

$$\delta = \sqrt{\frac{2}{\omega\mu\sigma}}$$

Charge redistribution model can be derived from the continuity equation

$$\frac{\delta\rho}{\delta t} + \frac{\sigma}{\epsilon}\rho = 0$$

3 Capacitors and Inductors

Φ | Magnetic Flux

n | Turns per unit length

ρ_l | Charge per unit length

h | Wire separation from central axis

a | Wire radius

$$\hat{C} = \frac{2\pi\epsilon}{\log(\frac{2h}{a})}$$

Coaxial cable

$$\hat{L} = \frac{\mu}{2\pi} \log(\frac{b}{a})$$

$$\hat{C} = \frac{2\pi\epsilon}{\log(\frac{b}{a})}$$

Definition of Capacitance - charge separation per volt

$$C(v) = \frac{dQ(v)}{dv}$$

Assuming linearity from the above

$$C = \frac{\epsilon \iiint_v \nabla \cdot \mathbf{E} dv}{\int_l \mathbf{E} \cdot d\mathbf{l}}$$

Definition of Inductance

$$Li = \Phi$$

Self Inductance

$$L(i) = \frac{d\Phi}{di}$$

Coil Inductance

$$\hat{L} = \mu n^2 A$$

Infinite parallel wire capacitance

$$\hat{C} = \frac{\pi\epsilon}{\log(\frac{h}{a} + \sqrt{(\frac{h}{a})^2 - 1})}$$

Two parallel wires

$$V = \frac{\rho_l}{\pi\epsilon} \log(\frac{h}{a} + \sqrt{(\frac{h}{a})^2 - 1})$$

$$\hat{L} = \frac{\mu}{\pi} \log(\frac{h-a}{a})$$

$$\hat{C} = \frac{\pi\epsilon}{\log(\frac{h}{a})}$$

Wire above plane

$$\hat{L} = \frac{\mu}{2\pi} \log(\frac{2h}{a})$$

4 Wires

Non-ideal properties

- Signal Distortion
 - Ringing response of conductors at high frequencies is due to inherent inductance and capacitance of wires (skin depth)
 - Wires act as a low pass filter
 - Affected by resistance, self-inductance and capacitance
 - Mitigated by
 - * Slower rise times, meaning a lower knee frequency so $f_{knee} \propto \frac{1}{\tau}$
 - * Shorter wire lengths, shifts the resonant behaviour to higher frequencies
 - * Flattening high frequency impedance (achieved through zero reflection coefficient)
- EMI
 - Electromagnetic radiation caused by quickly changing currents and voltages through conductors in high-speed digital systems
 - Amperes Law broadcast your digital signal wirelessly by accident
 - Caused by mutual inductance and capacitance
 - Big problem when there are current loops
 - * Current flowing out a pin, along a wire, into another device, and back via a ground plane or wire
 - * The area enclosed by the loop is the problem
 - * Behave like antennas
 - Mitigated by
 - * Increasing rise time
 - * Keep current loops in small area e.g coax cable
 - * Shielding
- Cross Talk
 - Induced voltages and currents due to EMI, causes noise
 - Faradays Law receive digital signals wirelessly by accident

- Caused by mutual inductance and capacitance
- Increased by a faster rise time, both in inductive and capacitive cross-talk
- Mitigated in same way as EMI

Consequences

- Ground bounce
 - Connection between internal and system ground is inductive, so due to $L = \frac{di}{dt}$ a large change in current will change ground voltage
 - Mitigated by
 - * Lower inductance packaging
 - * Larger diameter ground wires
 - * Ground wires closer to ground plane
 - * Edge slowing
 - * Lower voltage family
 - * Separate input group reference
 - * More ground wires

- h | Separation from center
- a | Smaller radius
- b | Larger radius
- ν | Signal propagation velocity
- D | Propagation delay per unit length
- t_{sw} | Switching time

Two parallel wires

$$\hat{L} = \frac{\mu}{\pi} \log\left(\frac{h-a}{a}\right)$$

$$\hat{C} = \frac{\pi\epsilon}{\log\left(\frac{h}{a}\right)}$$

Wire above plane

$$\hat{L} = \frac{\mu}{2\pi} \log\left(\frac{2h}{a}\right)$$

$$\hat{C} = \frac{2\pi\epsilon}{\log\left(\frac{2h}{a}\right)}$$

Coaxial cable

$$\hat{L} = \frac{\mu}{2\pi} \log\left(\frac{b}{a}\right)$$

$$\hat{C} = \frac{2\pi\epsilon}{\log\left(\frac{b}{a}\right)}$$

Signal propagation velocity

$$\nu = \frac{1}{\sqrt{\hat{L}\hat{C}}} = \frac{1}{\sqrt{\mu\epsilon}}$$

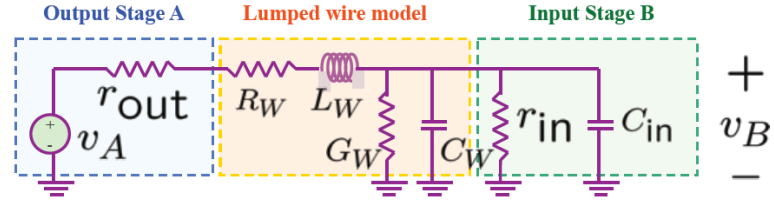
Propagation delay per unit length

$$D = \frac{1}{\nu} = \sqrt{\hat{L}\hat{C}} = \sqrt{\mu\epsilon}$$

Rising edge length

$$l_{sw} = \frac{t_{sw}}{D}$$

4.a Lumped Model



Y | Voltage over inductor

v_B | Voltage on receiving system side of capacitance

Lumped model occurs when $l_{sw} \gg l$
Knee frequency characterizes the approximate bandwidth of digital signals

$$f_{KNEE} = \frac{1}{\pi t_{sw}} \approx \frac{0.5}{t_{sw}}$$

Interconnect resonant frequency, signal will be distorted as the knee frequency exceeds the ringing frequency

$$f_{RING} = \frac{1}{2\pi} \sqrt{\frac{1}{LC} - \frac{R^2}{2L^2}} \approx \frac{1}{2\pi\sqrt{LC}}$$

Inductive Cross Talk

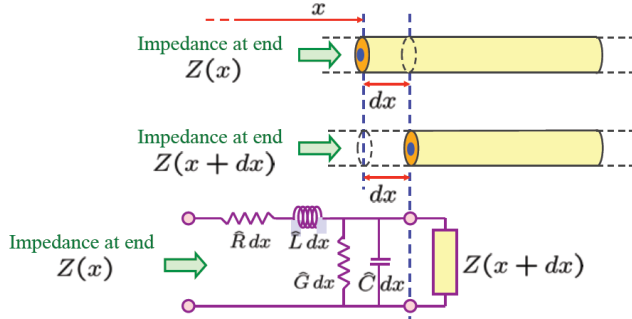
$$\frac{Y}{\Delta V} \approx \frac{L_M}{R_A t_{sw}} \approx \frac{L_M C_A}{t_{sw}^2}$$

Capacitive Cross Talk

$$\frac{v_B}{\Delta V} \approx \frac{R_B C_M}{t_{sw}}$$

4.b Distributed Model

G		Conductivity
ρ		Reflection Coefficient
Z_O		Characteristic Impedance
Γ		Attenuation Coefficient



Distributed model occurs when $l_{SW} \ll l$, these types of systems often suffer from signal distortion

$$\hat{Z}_{series} = \hat{R} + s\hat{L}$$

$$\hat{Y}_{shunt} = \hat{G} + s\hat{C}$$

Characteristic impedance is of an infinite length of wire

$$Z_O = \sqrt{\frac{\hat{Z}_{series}}{\hat{Y}_{shunt}}}$$

At low frequencies

$$Z_O(j\omega) = \sqrt{\frac{\hat{R}}{\hat{G}}}$$

At high frequencies (also the characteristic impedance of a lossless line)

$$Z_O(j\omega) = \sqrt{\frac{\hat{L}}{\hat{C}}}$$

Distributed impedance model for a finite length load terminated wire

$$\frac{dZ(x)}{dx} = \hat{Y}_{shunt}[Z(x)]^2 - \hat{Z}_{series}$$

$$Z(l) = Z_{load}$$

Resonant frequency

$$f_{ring} = \frac{1}{(4l)D}$$

Reflection Coefficient

$$\rho(s) = \frac{Z_{load}(s) - Z_O(s)}{Z_{load}(s) + Z_O(s)}$$

Attenuation Coefficient

$$\Gamma(s) = \sqrt{Z_{series}(s)Y_{shunt}(s)} = D_{lossless}s = s/v$$

Impedance for a finite length lossless wire

$$Z_{wire}(s) = Z_O(s) \left(\frac{1 + \rho e^{-2sl/v}}{1 - \rho e^{-2sl/v}} \right)$$

Transfer function for output voltage/input voltage

$$H_{AB}(s) = \frac{(1 + \rho_B \exp(-sl/v))}{1 + \rho_B \exp(-2sl/v)}$$

5 Semiconductors

Band-gap model

- Pauli's exclusion principle means no two identical particles can occupy the same quantum state
- Conduction band contains free to move electron
- Valence band contains electrons bound to specific atoms (immobile)
- Band gap is higher for insulators, lower for conductors
- Extrinsic semiconductors are intrinsic semiconductors with a dopant added
 - n-type is with a donor level added below conduction band that makes it easier for electrons to jump up (Group V dopant)
 - p-type is with an acceptor level added above the valence band that makes it easier for holes to move in the valence band (mediated by electrons)

n | Mobile conduction band electrons

N_A | Immobile acceptor ions

p | Mobile valence band holes

N_D | Immobile donor ions

T | Temperature

α | Recombination proportionality constant

Charge density is given by

$$\rho = q(p + N_D - n - N_A)$$

Electrons and holes are in pairs (dependent on temperature), so the EHP generation rate is

$$r_{gen} = n_i(T)^2$$

Electrons and holes also recombine in pairs, so the EHP recombination rate is dependent on concentration of both particles

$$r_{rec} = np$$

Electrons and holes can also be injected in via diffusion or contact with another conductor, therefore the mobile electron concentration model is

$$\frac{dn(t)}{dt} = \alpha[n_i^2 - n(t)p(t)] + r_{in}(t)$$

Under equilibrium conditions with no injection

$$\frac{dn(t)}{dt} = 0 = \frac{dp(t)}{dt}$$

$$\bar{n}\bar{p} = n_i^2$$

$$\bar{n} - \bar{p} = N_D - N_A$$

Therefore the concentrations are

$$\bar{n} = \frac{1}{2}(N_D - N_A) + \sqrt{\frac{1}{4}(N_A - N_D)^2 + n_i(T)^2}$$

For an n-type semiconductor

$$N_D \gg n_i^2, N_A = 0$$

$$\bar{n} \approx N_D$$

$$\bar{p} \approx \frac{n_i^2}{N_D}$$

For a p-type semiconductor

$$N_A \gg n_i^2, N_D = 0$$

$$\bar{p} \approx N_A$$

$$\bar{n} \approx \frac{n_i^2}{N_A}$$

5.a Currents in semiconductors

L_p	Diffusion length for holes
L_n	Diffusion length for electrons
D_p	Diffusion coefficient for holes
D_n	Diffusion coefficient for electrons
Δp	Hole perturbation at injection
Δn	Electron perturbation at injection

Current density in the semiconductor is due to drift and diffusion

$$\mathbf{J}_n = qn\mu_n\mathbf{E} + qD_n\nabla n$$

$$\mathbf{J}_p = qn\mu_p\mathbf{E} + qD_p\nabla p$$

$$\mathbf{J} = \mathbf{J}_n + \mathbf{J}_p$$

Where diffusion coefficients are

$$D_p = \frac{kT}{q}\mu_p$$

$$D_n = \frac{kT}{q}\mu_n$$

Recombination lifetime

$$\tau_r = \frac{1}{\alpha(\bar{n} + \bar{p})}$$

For p-type and n-type respectively, due to dominant carriers, the recombination rate is

$$\tau_r \approx \tau_p = \frac{1}{\alpha\bar{p}}$$

$$\tau_r \approx \tau_n = \frac{1}{\alpha\bar{n}}$$

If space charge neutrality is assumed, a perturbation in one carrier will instantly result in a perturbation in the other carrier, this is due to the 'instant' charge redistribution time compared to recombination, therefore

$$\rho = q(p + N_D - n - N_A) \text{ (from before)}$$

$$p(t) = n(t) + N_A - N_D$$

Define perturbations in the carriers as

$$\delta n(t, x) = n(t, x) - \bar{n}$$

$$\delta p(t, x) = p(t, x) - \bar{p}$$

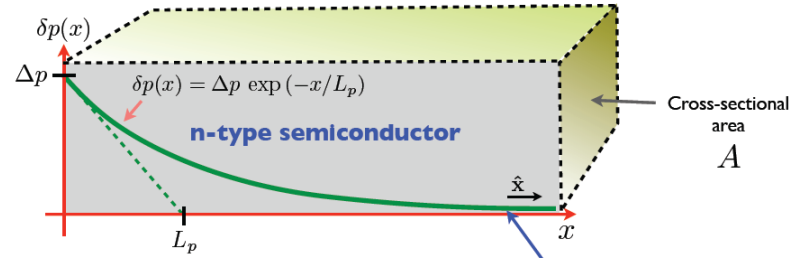
Yields a pair of continuity equations

$$\frac{1}{q}\nabla \cdot \mathbf{J}_p = -\frac{d\delta p}{dt} - \frac{\delta p}{\tau_p}$$

$$-\frac{1}{q}\nabla \cdot \mathbf{J}_n = -\frac{d\delta n}{dt} - \frac{\delta n}{\tau_n}$$

For the special case when there is no drift current, and all current is due to carrier diffusion, the diffusion current density due to electrons is

$$\mathbf{J}_n = qD_n\nabla n$$



The diffusion equations for electrons and holes are

$$\frac{d\delta n}{dt} = D_n\nabla^2(\delta n) - \frac{\delta n}{\tau_n}$$

$$\frac{d\delta p}{dt} = D_p\nabla^2(\delta p) - \frac{\delta p}{\tau_p}$$

Define the diffusion length of carriers as

$$L_p = \sqrt{D_p\tau_p}$$

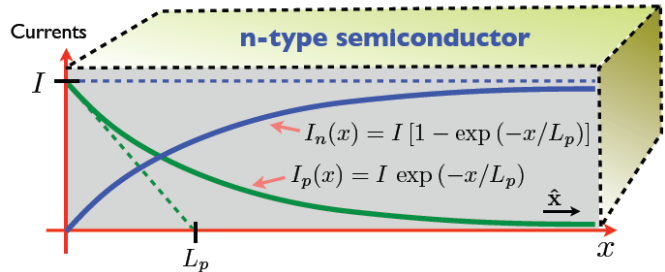
$$L_n = \sqrt{D_n\tau_n}$$

By solving with boundary conditions the excess carrier concentrations δp , δn can be found

$$\delta p(0) = \Delta p$$

$$\delta p(\infty) = 0$$

$$\delta p(x) = \Delta p \exp(-x/L_p)$$



The total current at any point flowing through the device is due to hole injection at $x = 0$, where electron current is zero

$$I = I_p(0) = \left(\frac{qAD_p}{L_p}\Delta p\right)$$

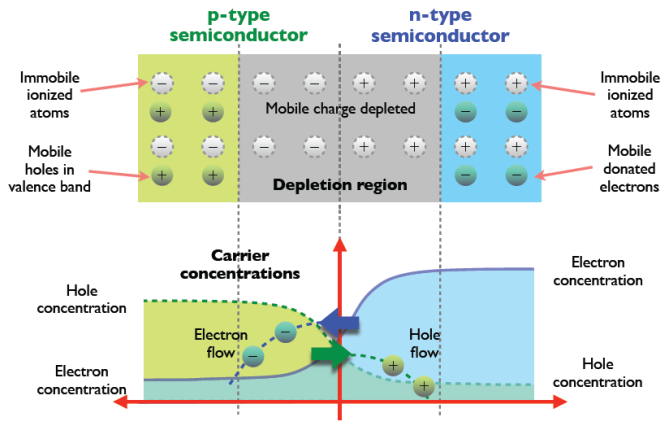
The electron current must pick up as hole current decays, the individual currents are

$$I_n(x) = I[1 - \exp(-x/L_p)]$$

$$I_p(x) = I \exp(-x/L_p)$$

$$I = I_n + I_p$$

5.b Junctions



- Carrier gradients between the two semiconductors leads to diffusion between the two
- Electrons diffuse from n to p
- Holes diffuse from p to n
- Mobile charge is depleted in central 'depletion region'
- Depletion approximation assumes that the entire voltage drop is over the depletion region
- Charge separation causes an electric field that opposes diffusion and creates equilibrium

- D_p | Diffusion coefficient for holes
- D_n | Diffusion coefficient for electrons
- ∇p | Gradient of p
- ∇n | Gradient of n
- μ_p | Hole mobility
- μ_n | Electron mobility
- p | Mobile hole concentration
- n | Mobile electron concentration
- k | Boltzmann's constant
- L_p | Diffusion length for holes
- L_n | Diffusion length for electrons
- Δp | Hole perturbation at injection
- Δn | Electron perturbation at injection

At equilibrium, $J_n = J_p = 0$

$$0 = qn\mu_n E + qD_n \nabla n$$

$$0 = qn\mu_p E + qD_p \nabla p$$

Therefore the electric field at equilibrium is

$$E = \frac{D_p}{\mu_p} \left(\frac{1}{p} \nabla p\right)$$

The "Einstein relation"

$$\frac{D_p}{\mu_p} = \frac{kT}{q}$$

Integrate over a path through this electric field with initial conditions of the injected carrier concentrations to find contact potential

$$V_O = \frac{D_p}{\mu_p} \log\left(\frac{p_p}{p_n}\right) = \frac{kT}{q} \log\left(\frac{p_p}{p_n}\right) \approx \frac{kT}{q} \log\left(\frac{N_A N_D}{n_i^2}\right)$$

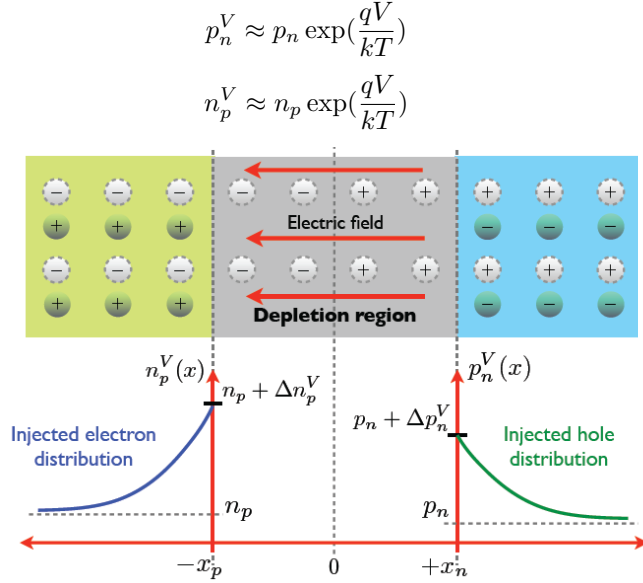
With an applied voltage V

$$V_O - V = \frac{D_p}{\mu_p} \log\left(\frac{p_p^V}{p_n^V}\right)$$

Solving for the new carrier concentrations

$$\frac{p_n^V p_p^V}{p_p^V p_n^V} = \exp\left(\frac{qV}{kT}\right)$$

$$p_p^V \approx p_p$$



Applying this to solve for the injected carrier distribution

$$\Delta p_n^V = p_n \left(\exp\left(\frac{qV}{kT}\right) - 1 \right)$$

$$\Delta n_p^V = n_p \left(\exp\left(\frac{qV}{kT}\right) - 1 \right)$$

Integrate to find the diffusion currents due to injected hole/electron distributions in n/p-type materials

$$I_p(x) = \left(\frac{qAD_p}{L_p} \right) \Delta p_n^V \exp\left(-\frac{x-x_n}{L_p}\right), \quad x \geq x_n$$

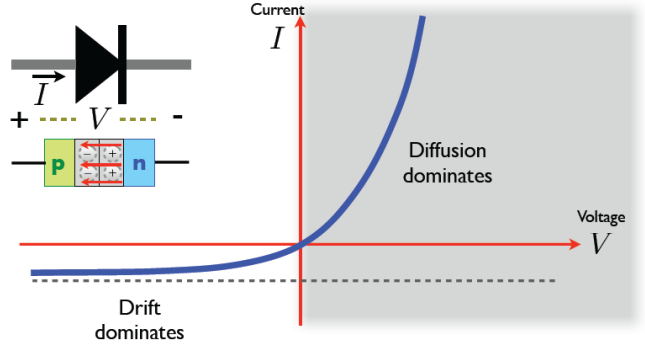
$$I_n(x) = \left(\frac{qAD_n}{L_n} \right) \Delta n_p^V \exp\left(-\frac{x+x_p}{L_n}\right), \quad x \geq x_n$$

Shockley Diode Equation can be obtained by summing the two currents

$$I = I_O \left[\exp\left(\frac{qV}{kT}\right) - 1 \right]$$

The saturation current is

$$I_O = qA \left(\frac{D_p p_n}{L_p} + \frac{D_n n_p}{L_n} \right)$$



Forward bias (diffusion dominates)

$$V \gg 0$$

$$I \approx I_O \exp\left(\frac{qV}{kT}\right)$$

Reverse bias (drift dominates)

$$V \ll 0$$

$$I \approx -I_O$$

Conductivity of neutral regions can be used to determine if negligible voltage drop assumption is valid

$$\sigma_p = q(\mu_p \bar{p}_p + \mu_n \bar{n}_p) \approx q\mu_p \bar{p}_p$$

$$\sigma_n = q(\mu_p \bar{p}_n + \mu_n \bar{n}_n) \approx q\mu_n \bar{n}_n$$

General Assumptions

- Electric field is confined to the junction and there is no electric field in neutral regions
- One dimensional device

5.c Junction Dynamics

There are two charge storage mechanism in junctions, and therefore two capacitances

- Depletion region capacitance
- Diffusion capacitance

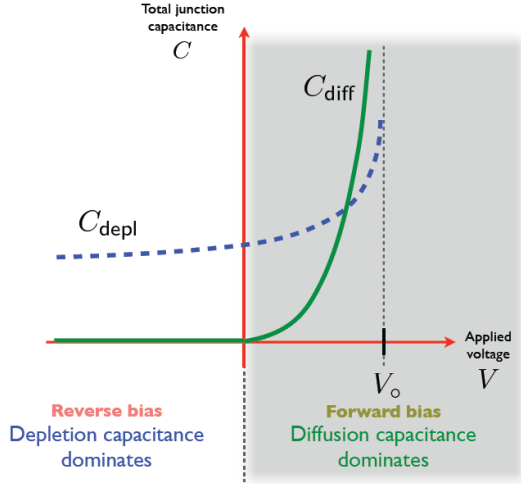
D_p | Diffusion coefficient for holes

p_n | Equilibrium minority carrier concentration (holes)

k | Boltzmann's constant

L_p | Diffusion length for holes

τ_p | Recombination lifetime (holes)



Depletion capacitance

$$q(t) = Q_{depl}(v(t))$$

$$C_{depl}(v) = \frac{dQ_{depl}}{dv}$$

$$C_{depl}(V) = \epsilon A \sqrt{\frac{q}{2\epsilon(V_0 - V) \frac{N_D N_A}{N_D + N_A}}}$$

Diffusion capacitance (two expressions for mobile holes and mobile electrons)

$$q(t) = Q_{diff}(v(t))$$

$$C_{diff}(v) = \frac{dQ_{diff}}{dv}$$

$$C_{depl}(V) = \frac{q^2}{kT} AL_p p_n \exp\left(\frac{qV}{kT}\right)$$

For a long p+n junction diode, we can assume junction operation is dominated by hole injection, and that all recombination happens before the end of the n-type material.

The total charge due to injected holes on the n-type region is:

$$Q_p(t) = \int_{x_n}^X qA \delta p_n(t, x) dx$$

Using this, the charge control model for both forward and reverse biased junctions can be derived

$$i(t) = \frac{dQ(t)}{dt} + \frac{Q(t)}{\tau}$$

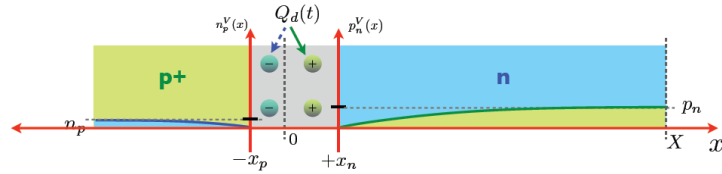
The charge storage delay for $p^+ - n$ and $p - n^+$ respectively

$$t_{CSD} = \tau_{p_n} \log\left(1 + \frac{I_f}{I_r}\right)$$

$$t_{CSD} = \tau_{n_p} \log\left(1 + \frac{I_f}{I_r}\right)$$

The charge control model can be applied for forward or reverse biased junctions.

- In forward bias, the charge is due to carrier injection in neutral regions either side of the depletion range. Separate models hold for both electrons and holes.
- In reverse bias, the charge is due to the "uncovered" ions in the depletion region



- Derivative term implies that changes in stored charge lag behind changes in current
- When the diode is "turned off"
 - Charge storage delay
 - * The applied voltage switches sign
 - * The excess of injected holes decays
 - * Diode voltage drops from contact potential down to zero
 - Carrier depletion
 - * Diode enters reverse bias
 - * Depletion region expands
 - * Diode voltage settles at $-E$
- For "turn on" the reverse
- Assumptions
 - n-type region is long $X \gg L_p$
 - p-type region is heavily doped compared with n-type region $p_n \gg n_p$
 - Half-period is much longer than recombination lifetime $T \gg \tau_p$
 - Forward bias junction voltage is limited by contact potential $v(t) \leq V_0$
 - Contact potential is much smaller than magnitude of applied voltage $V_0 \ll E$
 - Reverse bias saturation is very small $I_0 \approx 0$

6 Bipolar Junction Transistors

Applied Voltages	B-E junction bias	B-C junction bias	Mode
$E < B < C$	Forward	Reverse	Forward-active
$E < B > C$	Forward	Forward	Saturation
$E > B < C$	Reverse	Reverse	Cut-off
$E > B > C$	Reverse	Forward	Reverse-active

- Forward Active

1. Carrier injection - forward biased BE junction

- Injection of minority electrons into the p-type base region
- Injection of minority holes into the n-type emitter region

2. Carrier diffusion - base transport

- Diffusion of injected base electrons towards the collector
- Recombination of some injected electrons with majority holes
- Diffusion of remainder into the BC junction depletion region

3. Carrier Drift - BC depletion region

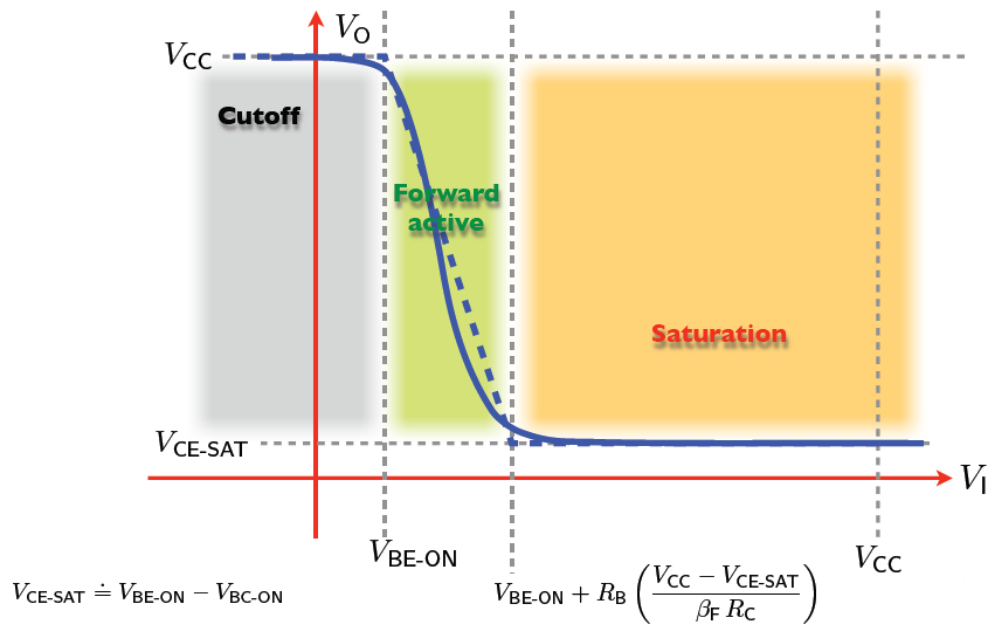
- Drift of minority electrons from base to collector
- Conventional current flows into the collector

- Large current gain due to minority electron current dominating BE current (BE junction is pn+)
- Collector current smaller than emitter current due to recombination across base and emitter hole current

- Reverse Active

- Saturation

- Cut-off



β | Current Gain

Ebers-Moll model
Current transfer ratios

$$\alpha_R = \frac{B_R}{1 + \beta_R}$$

$$\alpha_F = \frac{B_F}{1 + \beta_F}$$

BE junction current

$$I_F = I_{ES}[\exp(\frac{qV_{BE}}{kT}) - 1]$$

BC junction current

$$I_R = I_{CS}[\exp(\frac{qV_{BC}}{kT}) - 1]$$

Total emitter current

$$I_E = I_F - \alpha_R I_R$$

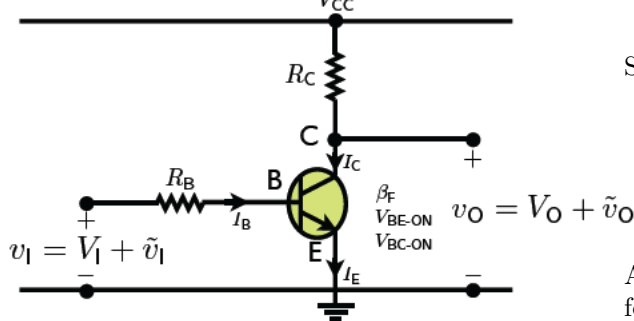
Total collector current

$$I_C = \alpha_F I_F - I_R$$

Total base current

$$I_B = I_E - I_C$$

Hybrid-Pi model



Local Gain

$$\frac{\tilde{v}_O}{\tilde{v}_I} \approx \frac{R_C \beta_F}{R_B + r_\pi}$$

6.a BJT Dynamics

Ebers-Moll and Hybrid-Pi are static models that ignore internal BJT dynamics, a dynamic model accounts for applied terminal voltages and currents that are time varying.

Steady state diffusion equation for electrons:

- Forward active
 - Approximately linear for narrow base widths
 - Charge injection is across BE junction only
 - Injected charge distribution dominates charge storage
- Cutoff
 - No injected charge
 - Charge stored in depletion region dipoles only
- Reverse active mode
 - Carrier injection across BC junction only
 - Injected charge dominates charge storage
- Saturation
 - Carrier injection across both junctions
 - Superposition of forward and reverse active modes
 - Injected charge dominates charge storage

$$\delta n(t, x) = n_p(t, x) - n_p$$

$$0 = D_N \frac{d^2 \delta n_p(x)}{dx^2} - \frac{\delta n_p(x)}{\tau_n}$$

Solve using boundary conditions

$$\delta n_p(0) = \Delta n_E = n_p[\exp(\frac{qV_{BE}}{kT}) - 1]$$

$$\delta n_p(W_b) = \Delta n_C = -n_p$$

Assuming that the p-type base material has relatively few electrons compared to that injected by the emitter, the excess electron concentration in the base is

$$\delta n_p(x) = \Delta n_p \left(\frac{\exp(\frac{W_b - x}{L_n}) - \exp(-\frac{W_b - x}{L_n})}{\exp(\frac{W_b}{L_n}) - \exp(-\frac{W_b}{L_n})} \right)$$

The terminal currents are therefore

$$I_E = qAD_n \frac{d\delta n_p(x)}{dx} \Big|_{x=0}$$

$$I_C = qAD_n \frac{d\delta n_p(x)}{dx} \Big|_{x=0}$$

$$I_B = I_E - I_C$$

- q_F | Charge stored in forward active mode excess carrier distribution
 q_R | Charge stored in reverse active mode excess carrier distribution
 q_{BE} | Charge stored in BE junction depletion region
 q_{BC} | Charge stored in BC junction depletion region
 τ_F | Mean minority carrier transit time across base in forward active mode
 τ_R | Mean minority carrier transit time across base in reverse active mode
 τ_{BF} | Minority carrier lifetime (in base) in forward active mode
 τ_{BR} | Minority carrier lifetime (in base) in reverse active mode

i_C	=	$-\frac{q_F}{\tau_F}$	$+\frac{dq_{BC}}{dt}$	$+q_R \left(\frac{1}{\tau_R} + \frac{1}{\tau_{BR}} \right) + \frac{dq_R}{dt}$
i_B	=	$-\frac{q_F}{\tau_{BF}} - \frac{dq_F}{dt}$	$-\frac{dq_{BE}}{dt} - \frac{dq_{BC}}{dt}$	$-\frac{q_R}{\tau_{BR}} - \frac{dq_R}{dt}$
i_E	=	$-q_F \left(\frac{1}{\tau_F} + \frac{1}{\tau_{BF}} \right) - \frac{dq_F}{dt}$	$-\frac{dq_{BE}}{dt}$	$+\frac{q_R}{\tau_R}$
		Electron injection across BE junction	Depletion region capacitances	Electron injection across BC junction